SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

 Eight 8-Bit Voltage Output DACs 5-V Single-Supply Operation 	N OR DW PACKAGE (TOP VIEW)
Serial Interface	
High-Impedance Reference Inputs	
 Programmable 1 or 2 Times Output Range 	GND 🛛 3 14 🗍 REF1
Simultaneous Update Facility	DATA 🛛 4 13 🗍 LDAC
Internal Power-On Reset	
Low-Power Consumption	V _{DD} [] 6 11 [] REF2
Half-Buffered Output	DACE [] 7 10]] DACH DACF [] 8 9]] DACG
applications	

applications

- Programmable Voltage Sources
- Digitally Controlled Amplifiers/Attenuators
- Mobile Communications
- Automatic Test Equipment
- Process Monitoring and Control
- Signal Synthesis

description

The TLC5628C and TLC5628I are octal 8-bit voltage output digital-to-analog converters (DACs) with buffered reference inputs (high impedance). The DACs produce an output voltage that ranges between either one or two times the reference voltages and GND and are monotonic. The device is simple to use, running from a single supply of 5 V. A power-on reset function is incorporated to ensure repeatable start-up conditions.

Digital control of the TLC5628C and TLC5628I are over a simple three-wire serial bus that is CMOS compatible and easily interfaced to all popular microprocessor and microcontroller devices. The 12-bit command word comprises eight bits of data, three DAC select bits, and a range bit, the latter allowing selection between the times 1 or times 2 output range. The DAC registers are double buffered, allowing a complete set of new values to be written to the device, then all DAC outputs are updated simultaneously through control of LDAC. The digital inputs feature Schmitt triggers for high-noise immunity.

The 16-terminal small-outline (D) package allows digital control of analog functions in space-critical applications. The TLC5628C is characterized for operation from 0°C to 70°C. The TLC5628I is characterized for operation from –40°C to 85°C. The TLC5628C and TLC5628I do not require external trimming.

PACKAGE										
TA	SMALL OUTLINE (DW)	PLASTIC DIP (N)								
0°C to 70°C	TLC5628CDW	TLC5628CN								
-40°C to 85°C	TLC5628IDW	TLC5628IN								

AVAILABLE OPTIONS



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SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

functional block diagram



Terminal Functions

TERMIN	IAL	1/0	DESCRIPTION
NAME	NO.	1/0	DESCRIPTION
CLK	5	I	Serial interface clock. The input digital data is shifted into the serial interface register on the falling edge of the clock applied to the CLK terminal.
DACA	2	0	DAC A analog output
DACB	1	0	DAC B analog output
DACC	16	0	DAC C analog output
DACD	15	0	DAC D analog output
DACE	7	0	DAC E analog output
DACF	8	0	DAC F analog output
DACG	9	0	DAC G analog output
DACH	10	0	DAC H analog output
DATA	4	I	Serial interface digital data input. The digital code for the DAC is clocked into the serial interface register serially. Each data bit is clocked into the register on the falling edge of the clock signal.
GND	3	I	Ground return and reference terminal
LDAC	13	I	Load DAC. When LDAC is high, no DAC output updates occur when the input digital data is read into the serial interface. The DAC outputs are only updated when LDAC is taken from high to low.
LOAD	12	I	Serial interface load control. When LDAC is low, the falling edge of the LOAD signal latches the digital data into the output latch and immediately produces the analog voltage at the DAC output terminal.
REF1	14	I	Reference voltage input to DAC A B C D. This voltage defines the analog output range.
REF2	11	I	Reference voltage input to DAC E F G H. This voltage defines the analog output range.
V _{DD}	6	I	Positive supply voltage



detailed description

The TLC5628 is implemented using eight resistor-string DACs. The core of each DAC is a single resistor with 256 taps, corresponding to the 256 possible codes listed in Table 1. One end of each resistor string is connected to GND and the other end is fed from the output of the reference input buffer. Monotonicity is maintained by use of the resistor strings. Linearity depends upon the matching of the resistor segments and upon the performance of the output buffer. Since the inputs are buffered, the DACs always present a high-impedance load to the reference sources. There are two input reference terminals; REF1 is used for DACA through DACD and REF2 is used by DACE through DACH.

Each DAC output is buffered by a configurable-gain output amplifier, that can be programmed to times 1 or times 2 gain.

On power up, the DACs are reset to CODE 0.

Each output voltage is given by:

 $V_{O}(DACA|B|C|D|E|F|G|H) = REF \times \frac{CODE}{256} \times (1 + RNG \text{ bit value})$

where CODE is in the range 0 to 255 and the range (RNG) bit is a 0 or 1 within the serial control word.

D7	D6	D5	D4	D3	D2	D1	D0	OUTPUT VOLTAGE
0	0	0	0	0	0	0	0	GND
0	0	0	0	0	0	0	1	(1/256) × REF (1+RNG)
•	•	•	•	•	•	•	•	•
•	•	•	•	•	•	•	•	•
0	1	1	1	1	1	1	1	(127/256) × REF (1+RNG)
1	0	0	0	0	0	0	0	(128/256) × REF (1+RNG)
•	•	•	•	•	•	•	•	•
•	•	•	•	•	•	•	•	•
1	1	1	1	1	1	1	1	(255/256) × REF (1+RNG)

Table 1. Ideal Output Transfer

data interface

With LOAD high, data is clocked into the DATA terminal on each falling edge of CLK. Once all data bits have been clocked in, LOAD is pulsed low to transfer the data from the serial input register to the selected DAC as shown in Figure 1. When LDAC is low, the selected DAC output voltage is updated when LOAD goes low. When LDAC is high during serial programming, the new value is stored within the device and can be transferred to the DAC output at a later time by pulsing LDAC low as shown in Figure 2. Data is entered most significant bit (MSB) first. Data transfers using two 8-clock cycle periods are shown in Figures 3 and 4.







SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997



Figure 4. LDAC-Controlled Update Using 8-Bit Serial Word

Table 2 lists the A2, A1, and A0 bits and the selection of the updated DACs. The RNG bit controls the DAC output range. When RNG = low, the output range is between the applied reference voltage and GND, and when RNG = high, the range is between twice the applied reference voltage and GND.

A2	A1	A0	DAC UPDATED
0	0	0	DACA
0	0	1	DACB
0	1	0	DACC
0	1	1	DACD
1	0	0	DACE
1	0	1	DACF
1	1	0	DACG
1	1	1	DACH

Table 2. Serial Input Decode



linearity, offset, and gain error using single-end supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset voltage, the output voltage changes on the first code change. With a negative offset the output voltage may not change with the first code depending on the magnitude of the offset voltage.

The output amplifier, therefore, attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground.

The output voltage remains at 0 V until the input code value produces a sufficient output voltage to overcome the inherent negative offset voltage, resulting in the transfer function shown in Figure 5.



Figure 5. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces the breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below ground.

For a DAC, linearity is measured between the zero-input code (all inputs are 0) and the full-scale code (all inputs are 1) after offset and full scale are adjusted out or accounted for in some way. However, single-supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity in the unipolar mode is measured between full-scale code and the lowest code that produces a positive output voltage.

The code is calculated from the maximum specification for the negative offset voltage.



SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

equivalent of inputs and outputs



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Supply voltage (V _{DD} – GND)	
Digital input voltage range, VID	
Reference input voltage range	GND – 0.3 V to V_{DD}^{-} + 0.3 V
Operating free-air temperature range, T _A : TLC5628C	$0^{\circ}\overline{C}$ to $70^{\circ}C$
TLC5628I	–40°C to 85°C
Storage temperature range, T _{stg}	–50°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V _{DD}		4.75		5.25	V
High-level voltage, VIH		0.8 V _{DD}			V
Low-level voltage, VIL				0.8	V
Reference voltage, Vref [A B C D E F G	4]			V _{DD} -1.5	V
Analog full-scale output voltage, $R_L = 10$	kΩ		3.5		V
Load resistance, RL		10	kΩ		
Setup time, data input, t _{SU(DATA-CLK)} (s	50			ns	
Valid time, data input valid after $CLK{\downarrow},t_V$	50			ns	
Setup time, CLK eleventh falling edge to	LOAD, t _{su(CLK-LOAD)} (see Figure 1)	50			ns
Setup time, LOAD \uparrow to CLK \downarrow , t _{SU} (LOAD-	CLK) (see Figure 1)	50			ns
Pulse duration, LOAD, $t_{W(LOAD)}$ (see Fi	gure 1)	250			ns
Pulse duration, LDAC, tw(LDAC) (see Fig	gure 2)	250			ns
Setup time, LOAD↑ to LDAC↓, t _{SU(LOAD}	o-LDAC) (see Figure 2)	0			ns
CLK frequency				1	MHz
Operating free-air temperature, T _A	TLC5628C	0		70	°C
	TLC5628I	-40		85	°C



SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

electrical characteristics over recommended operating free-air temperature range, V_{DD} = 5 V ± 5%,
V_{ref} = 2 V, \times 1 gain output range (unless otherwise noted)

	PARAMETER	TES	T CONDITIONS	MIN	TYP	MAX	UNIT
IIН	High-level input current	$V_I = V_{DD}$	$V_{I} = V_{DD}$			±10	μA
۱ _{IL}	Low-level input current	V _I = 0 V	$V_{I} = 0 V$			±10	μA
IO(sink)	Output sink current	Each DAC ou	itout	20			μA
IO(source)	Output source current	Each DAC OU	lipui	2			mA
C.	Input capacitance				15		~F
Ci	Reference input capacitance				15		pF
IDD	Supply current	V _{DD} = 5 V				4	mA
Iref	Reference input current	V _{DD} = 5 V,	V _{ref} = 2 V			±10	μA
EL	Linearity error (end point corrected)	V _{ref} = 2 V,	× 2 gain (see Note 1)			±1	LSB
ED	Differential-linearity error	V _{ref} = 2 V,	× 2 gain (see Note 2)			±0.9	LSB
E _{ZS}	Zero-scale error	V _{ref} = 2 V,	× 2 gain (see Note 3)	0		30	mV
	Zero-scale-error temperature coefficient	V _{ref} = 2 V,	× 2 gain (see Note 4)		10		μV/°C
E _{FS}	Full-scale error	V _{ref} = 2 V,	imes2 gain (see Note 5)			±60	mV
	Full-scale-error temperature coefficient	V _{ref} = 2 V,	× 2 gain (see Note 6)		±25		μV/°C
PSRR	Power supply rejection ratio	See Notes 7	See Notes 7 and 8		0.5		mV/V

NOTES: 1. Integral nonlinearity (INL) is the maximum deviation of the output from the line between zero and full scale (excluding the effects of zero code and full-scale errors).

2. Differential nonlinearity (DNL) is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

3. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.

4. Zero-scale-error temperature coefficient is given by: $ZSETC = [ZSE(T_{max}) - ZSE(T_{min})]/V_{ref} \times 10^{6}/(T_{max} - T_{min})$.

5. Full-scale error is the deviation from the ideal full-scale output ($V_{ref} - 1$ LSB) with an output load of 10 k Ω .

6. Full-scale error temperature coefficient is given by: FSETC = [FSE(T_{max}) - FSE (T_{min})]/V_{ref} × 10⁶/(T_{max} - T_{min}).

 Zero-scale-error rejection ratio (ZSE RR) is measured by varying the V_{DD} from 4.5 V to 5.5 V dc and measuring the proportion of this signal imposed on the zero-code output voltage.

8. Full-scale-error rejection ratio (FSE RR) is measured by varying the V_{DD} from 4.5 V to 5.5 V dc and measuring the proportion of this signal imposed on the full-scale output voltage.

operating characteristics over recommended operating free-air temperature range, V_{DD} = 5 V ± 5%, V_{ref} = 2 V, × 1 gain output range (unless otherwise noted)

	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output slew rate	$C_L = 100 \text{ pF}, \qquad R_L = 10 \text{ k}\Omega$		1		V/µs
Output settling time	To ± 0.5 LSB, C _L = 100 pF, R _L = 10 k Ω , See Note 9		10		μs
Large signal bandwidth	Measured at -3 dB point		100		kHz
Digital crosstalk	CLK = 1-MHz square wave measured at DACA-DACD		-50		dB
Reference feedthrough	See Note 10		-60		dB
Channel-to-channel isolation	See Note 11		-60		dB
Reference input bandwidth	See Note 12		100		kHz

NOTES: 9. Settling time is the time between a LOAD falling edge and the DAC output reaching full-scale voltage within ±0.5 LSB starting from an initial output voltage equal to zero.

10. Reference feedthrough is measured at any DAC output with an input code = 00 hex with a V_{ref} input = 1 V dc + 1 V_{pp} at 10 kHz. 11. Channel-to-channel isolation is measured by setting the input code of one DAC to FF hex and the code of all other DACs to 00 hex with V_{ref} input = 1 V dc + 1 V = at 10 kHz.

with V_{ref} input = 1 V dc + 1 V_{pp} at 10 kHz. 12. Reference bandwidth is the -3 dB bandwidth with an input at V_{ref} = 1.25 V dc + 2 V_{pp} and with a full-scale digital input code.



SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

PARAMETER MEASUREMENT INFORMATION





TYPICAL CHARACTERISTICS





SLAS089E – NOVEMBER 1994 – REVISED APRIL 1997

TYPICAL CHARACTERISTICS



SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997



TYPICAL CHARACTERISTICS

APPLICATION INFORMATION





Figure 15. Output Buffering Scheme



SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

MECHANICAL DATA

PLASTIC SMALL-OUTLINE PACKAGE

DW (R-PDSO-G**) 16 PIN SHOWN



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).

D. Falls within JEDEC MS-013



SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

MECHANICAL DATA

PLASTIC DUAL-IN-LINE PACKAGE

N (R-PDIP-T**)



- NOTES: A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 (20-pin package is shorter than MS-001)



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PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Packag Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TLC5628CDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628CDWG4	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628CDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628CDWRG4	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628CN	ACTIVE	PDIP	Ν	16	25	Pb-Free (RoHS)	CU NIPD	N / A for Pkg Type
TLC5628CNE4	ACTIVE	PDIP	Ν	16	25	Pb-Free (RoHS)	CU NIPD	N / A for Pkg Type
TLC5628IDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628IDWG4	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628IDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628IDWRG4	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC5628IN	ACTIVE	PDIP	Ν	16	25	Pb-Free (RoHS)	CU NIPD	N / A for Pkg Type
TLC5628INE4	ACTIVE	PDIP	Ν	16	25	Pb-Free (RoHS)	CU NIPD	N / A for Pkg Type

⁽¹⁾ The marketing status values are defined as follows:

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PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

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⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	-	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC5628CDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
TLC5628IDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1



PACKAGE MATERIALS INFORMATION

11-Mar-2008



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC5628CDWR	SOIC	DW	16	2000	346.0	346.0	33.0
TLC5628IDWR	SOIC	DW	16	2000	346.0	346.0	33.0

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