

## STEREO AUDIO AMPLIFIER

- DUAL OR BRIDGE CONNECTION MODES
- FEW EXTERNAL COMPONENTS
- SUPPLY VOLTAGE DOWN TO 3V
- HIGH CHANNEL SEPARATION
- VERY LOW SWITCH ON/OFF NOISE
- MAX GAIN OF 45dB WITH ADJUST EXTER-NAL RESISTOR
- SOFT CLIPPING
- THERMAL PROTECTION
- 3V < V<sub>CC</sub> < 15V
- $\bullet P = 2 \bullet 1W, V_{CC} = 6V, R_L = 4\Omega$
- $\blacksquare P = 2 \bullet 2.3W, V_{CC} = 9V, R_L = 4\Omega$
- $P = 2 \bullet 0.1W, V_{CC} = 3V, R_L = 4\Omega$



#### DESCRIPTION

The TEA2025B/D is a monolithic integrated circuit in 12+2+2 Powerdip and 12+4+4 SO, intended for use as dual or bridge power audio amplifier portable radio cassette players.

Symbol	Parameter	Test Conditions	Unit
Vs	Supply Voltage	15	V
lo	Ouput Peak Current	1.5	А
TJ	Junction Temperature	150	°C
T <sub>stg</sub>	Storage Temperature	150	°C

#### **BLOCK DIAGRAM**



## ABSOLUTE MAXIMUM RATINGS



SO 12+4+4 PIN CONNECTION (Top view)



#### THERMAL DATA

Symbol	Description	SO 12+4+4 (*)	PDIP 12+2+2 (**)	Unit	
R <sub>th j-case</sub>	Thermal Resistance Junction-case	Max	15	15	°C/W
R <sub>th j-amb</sub>	Thermal Resistance Junction-ambient	Max	65	60	°C/W

(\*) The  $R_{th\;j\text{-amb}}$  is measured with  $4sq\,cm$  copper area heatsink

(\*\*) The  $R_{th\,j\text{-amb}}$  is measured on devices bonded on a 10 x 5 x 0.15cm glass-epoxy substrate with a 35 $\mu$ m thick copper surface of 5 cm<sup>2</sup>.



Symbol	Parameter	Test Conditions	S		Min.	Тур.	Max.	Unit
Vs	Supply Voltage				3		12	V
lq	Quiescent Current					35	50	mA
Vo	Quiescent Output Voltage					4.5		V
Av	Voltage Gain	Stereo Bridge			43 49	45 51	47 53	dB
$\Delta A_V$	Voltage Gain Difference						±1	dB
Rj	Input Impedance					30		KΩ
PO	Output Power (d = 10%)	Stereo 8 (per channel)	9V 9V 6V 6V 6V 6V 3V 3V 2V	4Ω 8Ω 4Ω 8Ω 16Ω 32Ω 4Ω 32Ω 8Ω	1.7 0.7	2.3 1.3 1 0.6 0.25 0.13 0.1 0.02 2.4		w
		Bridge	9V 6V 6V 3V 3V	8Ω 4Ω 8Ω 16Ω 32Ω		4.7 2.8 1.5 0.18 0.06		W
d	Distortion	$Vs = 9V; R_L = 4\Omega$	Stereo Bridge			0.3 0.5	1.5	%
SVR	Supply Voltage Rejection	$f = 100Hz, V_R = 0.5V, R_g = 0$			40	46		dB
E <sub>N</sub> (IN)	Input Noise Voltage	$R_G = 0$ $R_G = 10 \ 4\Omega$				1.5 3	3 6	mV
СТ	Cross-Talk	f = 1KHz, R <sub>g</sub> = 10KΩ			40	52		dB

16 Term. N° (PDIP) 2 3 4 5 6 7 8 9 10 11 12 13 14 15 1 DC VOLT (V) 0.04 4.5 8.9 0 0 0.6 0.04 8.5 0 0.04 0.6 0 0 8.9 4.5 9

Figure 1: Bridge Application (Powerdip)



Figure 2: Stereo Application (Powerdip)





Figure 3: Supply Current vs. Supply Voltage  $(R_L = 4\Omega)$ 



Figure 4: Output Voltage vs. Supply Voltage



STEREO

15







Vo(V)

8

7 6

5 Δ

3 2

1

3.5

3

2.5

2

1.5

1

0.5

0 <sup>E</sup>

#### **APPLICATION INFORMATION**

#### **Input Capacitor**

Input capacitor is PNP type allowing source to be referenced to ground.

In this way no input coupling capacitor is required. However, a series capacitor (0.22 uF)to the input side can be useful in case of noise due to variable resistor contact.

#### Bootstrap

The bootstrap connection allows to increase the output swing.

The suggested value for the bootstrap capacitors (100uF) avoids a reduction of the output signal also at low frequencies and low supply voltages.

#### Voltage Gain Adjust

#### STEREO MODE

The voltage gain is determined by on-chip resistors R1 and R2 together with the external RfC1 series connected between pin 6 (11) and ground.

The frequency response is given approximated by:

$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf + R2 + \frac{1}{MAC}}$$

With Rf=0, C1=100 uF, the gain results 46 dB with pole at f=32 Hz.

THE purpose of Rf is to reduce the gain. It is recommended to not reduce it under 36 dB.

#### BRIDGE MODE

Figure 7



The bridge configuration is realized very easily thanks to an internal voltage divider which provides (at pin 1) the CH 1 output signal after reduction. It is enough to connect pin 6 (inverting input of CH 2) with a capacitor to pin 1 and to connect to ground the pin 7.

The total gain of the bridge is given by:

$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf + R2 + \frac{1}{R} (1 + \frac{R3}{R4} \frac{R1}{R2 + R4 + \frac{1}{R} (1 + \frac{R3}{R4} \frac{R1}{R2 + R4 + \frac{1}{R} (1 + \frac{R3}{R4} \frac{R1}{R4} \frac{R1}{R4})})$$

and with the suggested values (C1 = C2 =  $100 \ \mu$ F, Rf= 0) means:

Gv = 52 dB





with first pole at f = 32 Hz

#### **Output Capacitors.**

The low cut off frequency due to output capacitor depending on the load is given by:

$$F_L = \frac{1}{2 \Pi C_{OUT} \bullet R_L}$$

with  $C_{OUT}$  470 $\!\mu F$  and  $R_L$  = 4 ohm it means  $F_L$  = 80 Hz.

#### Pop Noise

Most amplifiers similar to TEA 2025B need external resistors between DC outputs and ground in order to optimize the pop on/off performance and crossover distortion.

#### Figure 9



The TEA 2025B solution allows to save components because of such resistors (800 ohm)are included into the chip.



#### Stability

A good layout is recommended in order to avoid oscillations.

Generally the designer must pay attention on the following points:

- Short wires of components and short connections.

- No ground loops.

- Bypass of supply voltage with capacitors as nearest as possible to the supply I.C.pin.The low value(poliester)capacitors must have good temperature and frequency characteristics.

#### - No sockets.

2) the heatsink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of excessive junction temperature: all that happens is that  $P_O$  (and therefore  $P_{tot}$ ) and Id are reduced.

#### **APPLICATION SUGGESTION**

The recommended values of the components are those shown on stereo application circuit of Fig. 2 different values can be used, the following table can help the designer.

COMPONENT	RECOMMENDED VALUE	PURPOSE	LARGER THAN	SMALLER THAN
C1,C2	0.22µF	INPUT DC DECOUPLING IN CASE OF SLIDER CONTACT NOISE OF VARIABLE RESISTOR		
C3	100µF	RIPPLE REJECTON		DEGRADATION OF SVR, INCREASE OF THD AT LOW FREQUENCY AND LOW VOLTAGE
C4,C5	100µF	BOOTSTRAP		
C6,C7	470µF	OUTPUT DC DECOUPLING		INCREASE OF LOW FREQUENCY CUT- OFF
C8,C9	0.15µF	FREQUENCY STABILITY		DANGER OF OSCILLATIONS
C10, C11	100µF	INVERTING INPUT DC DECOUPLING		INCREASE OF LOW FREQUENCY CUT- OFF



## SO20 PACKAGE MECHANICAL DATA

DIM.	mm			inch			
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
A			2.65			0.104	
a1	0.1		0.3	0.004		0.012	
a2			2.45			0.096	
b	0.35		0.49	0.014		0.019	
b1	0.23		0.32	0.009		0.013	
С		0.5			0.020		
c1			45 (	typ.)			
D	12.6		13.0	0.496		0.512	
E	10		10.65	0.394		0.419	
е		1.27			0.050		
e3		11.43			0.450		
F	7.4		7.6	0.291		0.299	
L	0.5		1.27	0.020		0.050	
м			0.75			0.030	
S	8 (max.)						



#### **DIP16 PACKAGE MECHANICAL DATA**

DIM.	mm			inch			
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
a1	0.51			0.020			
В	0.85		1.40	0.033		0.055	
b		0.50			0.020		
b1	0.38		0.50	0.015		0.020	
D			20.0			0.787	
E		8.80			0.346		
е		2.54			0.100		
e3		17.78			0.700		
F			7.10			0.280	
I			5.10			0.201	
L		3.30			0.130		
Z			1.27			0.050	





Information furnished is believed to be accurate and reliable. However, SGS-THOMSON Microelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SGS-THOMSON Microelectronics. Specifications mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. SGS-THOMSON Microelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of SGS-THOMSON Microelectronics.

#### © 1994 SGS-THOMSON Microelectronics - All Rights Reserved

SGS-THOMSON Microelectronics GROUP OF COMPANIES

Australia - Brazil - France - Germany - Hong Kong - Italy - Japan - Korea - Malaysia - Malta - Morocco - The Netherlands - Singapore - Spain - Sweden - Switzerland - Taiwan - Thaliand - United Kingdom - U.S.A.



This datasheet has been download from:

www.datasheetcatalog.com

Datasheets for electronics components.